



Dual P-Channel 12 V (D-S) MOSFET

PRODUCT SUMMARY						
V _{DS} (V)	$R_{DS(on)}(\Omega)$	I _D (A)	Q _g (Typ.)			
	0.390 at V _{GS} = - 4.5 V	- 1.3 ^a				
- 12	0.535 at V _{GS} = - 2.5 V	- 1.2	1.7 nC			
	0.710 at V _{GS} = - 1.8 V	- 1.1				

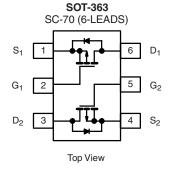
FEATURES

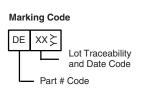
- Halogen-free According to IEC 61249-2-21 Definition
- TrenchFET[®] Power MOSFET
- Compliant to RoHS Directive 2002/95/EC

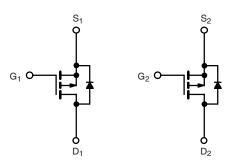


APPLICATIONS

· Load Switch for Portable Devices







Ordering Information: Si1965DH-T1-E3 (Lead (Pb)-free)

Si1965DH-T1-GE3 (Lead (Pb)-free and Halogen-free)

P-Channel MOSFET P-Channel MOSFET

ABSOLUTE MAXIMUM RATINGS T _A = 25 °C, unless otherwise noted					
Parameter	Symbol	Limit	Unit		
Drain-Source Voltage		V_{DS}	- 12	V	
Gate-Source Voltage		V _{GS}	± 8	v	
	T _C = 25 °C		- 1.3 ^a		
Continuous Drain Current /T 150 °C)	T _C = 70 °C		- 1.2		
Continuous Drain Current (T _J = 150 °C)	T _A = 25 °C	I _D	- 1.14 ^{b, c}		
	T _A = 70 °C		- 0.9 ^{b, c}	Α	
Pulsed Drain Current		I _{DM}	- 3		
	T _C = 25 °C		- 1		
Continuous Source-Drain Diode Current	T _A = 25 °C	I _S	- 0.6 ^{b, c}		
	T _C = 25 °C		1.25		
Maximum Power Dissipation	T _C = 70 °C	ь	0.8	w	
	T _A = 25 °C	- P _D	0.74 ^{b, c}	VV	
	T _A = 70 °C		0.47 ^{b, c}		
Operating Junction and Storage Temperature Range		T _J , T _{stg}	- 55 to 150	°C	

THERMAL RESISTANCE RATINGS							
Parameter	Symbol	Typical	Maximum	Unit			
Maximum Junction-to-Ambient ^{b, d}	t ≤ 5 s	R_{thJA}	130	170	°C/W		
Maximum Junction-to-Foot (Drain)	Steady State	R_{thJF}	80	100	O, VV		

Notes:

- a. Package limited.
- b. Surface mounted on 1" x 1" FR4 board.
- c. t = 5 s.
- d. Maximum under steady state conditions is 220 $^{\circ}\text{C/W}.$



Parameter	Symbol	Test Conditions	Min.	Тур.	Max.	Unit	
Static					L		
Drain-Source Breakdown Voltage	V _{DS}	V _{GS} = 0 V, I _D = - 250 μA	- 12			٧	
V _{DS} Temperature Coefficient	$\Delta V_{DS}/T_{J}$	$\Delta V_{DS}/T_J$		- 14			
V _{GS(th)} Temperature Coefficient	$\Delta V_{GS(th)}/T_J$			2		mV/°C	
Gate-Source Threshold Voltage	V _{GS(th)}	$V_{DS} = V_{GS}$, $I_{D} = -250 \mu A$	- 0.4		- 1.0	٧	
Gate-Source Leakage	I _{GSS}	$V_{DS} = 0 \text{ V}, V_{GS} = \pm 8 \text{ V}$			± 100	nA	
Zana Oata Walkana Busin Oamant		V _{DS} = - 12 V, V _{GS} = 0 V			- 1		
Zero Gate Voltage Drain Current	I _{DSS}	$V_{DS} = -12 \text{ V}, V_{GS} = 0 \text{ V}, T_{J} = 85 ^{\circ}\text{C}$			- 10	μΑ	
On-State Drain Current ^a	I _{D(on)}	$V_{DS} \le -5 \text{ V}, V_{GS} = -4.5 \text{ V}$	- 3			Α	
	, ,	$V_{GS} = -4.5 \text{ V}, I_D = -1.0 \text{ A}$		0.315	0.390		
Drain-Source On-State Resistance ^a	R _{DS(on)}	$V_{GS} = -2.5 \text{ V}, I_D = -0.86 \text{ A}$		0.425	0.535	Ω	
	, ,	V _{GS} = - 1.8 V, I _D = - 0.25 A		0.550	0.710	1	
Forward Transconductance ^a	9 _{fs}	V _{DS} = - 6 V, I _D = - 1.0 A		2.5		S	
Dynamic ^b				•			
Input Capacitance	C _{iss}			120		pF	
Output Capacitance	C _{oss}	V _{DS} = - 6 V, V _{GS} = 0 V, f = 1 MHz		41			
Reverse Transfer Capacitance	C _{rss}			25			
Total Gate Charge	Qg	$V_{DS} = -6 \text{ V}, V_{GS} = -8 \text{ V}, I_{D} = -1.1 \text{ A}$		2.8	4.2		
				1.7	2.6		
Gate-Source Charge	Q_{gs}	$V_{DS} = -6 \text{ V}, V_{GS} = -4.5 \text{ V}, I_{D} = -1.1 \text{ A}$		0.3		nC	
Gate-Drain Charge	Q _{gd}			0.4			
Gate Resistance	R_{g}	f = 1 MHz		7.5		Ω	
Turn-On Delay Time	t _{d(on)}			12	20		
Rise Time	t _r	$V_{DD} = -6 \text{ V}, R_1 = 6.7 \Omega$		27	40	1	
Turn-Off Delay Time	t _{d(off)}	$I_D \cong -0.9 \text{ A}, V_{GEN} = -4.5 \text{ V}, R_g = 1 \Omega$		15	25		
Fall Time	t _f			10	15		
Turn-On Delay Time	t _{d(on)}			2	5	ns	
Rise Time	t _r	V_{DD} = - 6 V, R_L = 6.7 Ω		12	20	1	
Turn-Off Delay Time	t _{d(off)}	$I_D \cong -0.9 \text{ A}, V_{GEN} = -8 \text{ V}, R_g = 1 \Omega$		12	20		
Fall Time	t _f			10	15		
Drain-Source Body Diode Characteristic	s			•	! 		
Continuous Source-Drain Diode Current	I _S	T _C = 25 °C			- 1.0		
Pulse Diode Forward Current ^a	I _{SM}				- 3.0	A	
Body Diode Voltage	V _{SD}	I _S = - 0.9 A		- 0.8	- 1.2	٧	
Body Diode Reverse Recovery Time	t _{rr}	-		20	40	ns	
Body Diode Reverse Recovery Charge	Q _{rr}			10	20	nC	
Reverse Recovery Fall Time	t _a	$I_F = -0.9 \text{ A}, \text{ dI/dt} = 100 \text{ A/}\mu\text{s}, T_J = 25 ^{\circ}\text{C}$		9.5			
Reverse Recovery Rise Time	t _b			11.5		ns	

Notes:

- a. Pulse test; pulse width \leq 300 μ s, duty cycle \leq 2 %.
- b. Guaranteed by design, not subject to production testing.

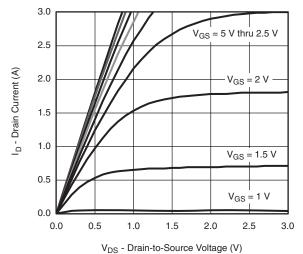
Stresses beyond those listed under "Absolute Maximum Ratings" may cause permanent damage to the device. These are stress ratings only, and functional operation of the device at these or any other conditions beyond those indicated in the operational sections of the specifications is not implied. Exposure to absolute maximum rating conditions for extended periods may affect device reliability.



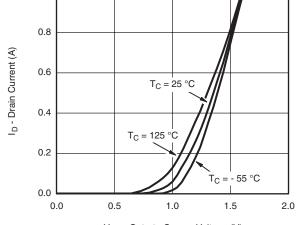




TYPICAL CHARACTERISTICS 25 °C, unless otherwise noted



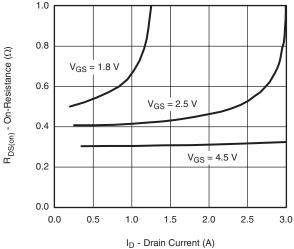
Output Characteristics



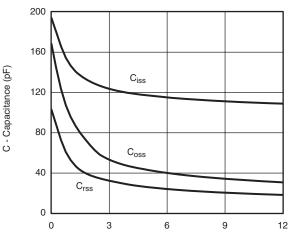
1.0

V_{GS} - Gate-to-Source Voltage (V) **Transfer Characteristics**

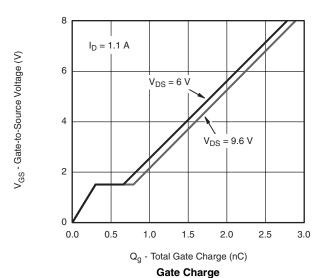


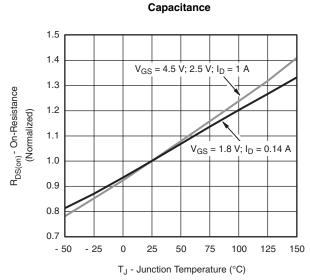


On-Resistance vs. Drain Current and Gate Voltage



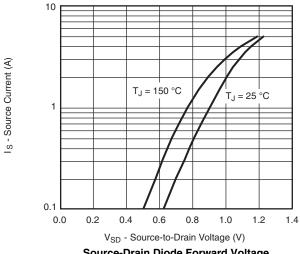
V_{DS} - Drain-to-Source Voltage (V)

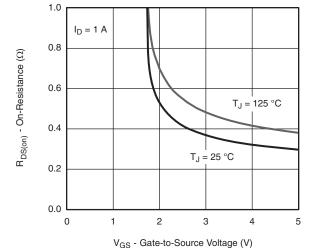




On-Resistance vs. Junction Temperature

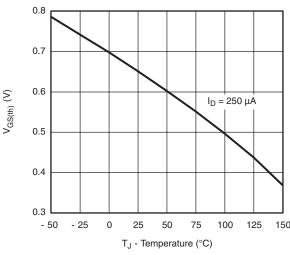
TYPICAL CHARACTERISTICS 25 °C, unless otherwise noted

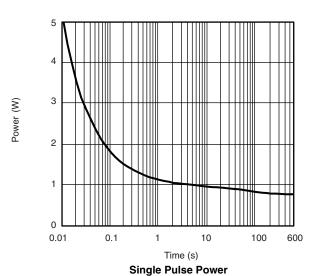




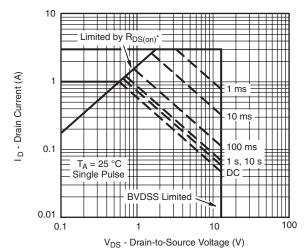
Source-Drain Diode Forward Voltage







Threshold Voltage



* V_{GS} > minimum V_{GS} at which $R_{DS(on)}$ is specified

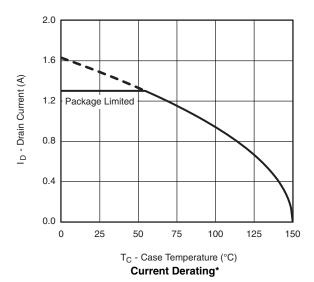
Safe Operating Area, Junction-to-Ambient

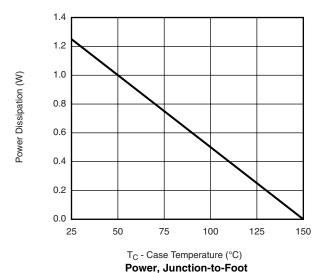






TYPICAL CHARACTERISTICS 25 °C, unless otherwise noted

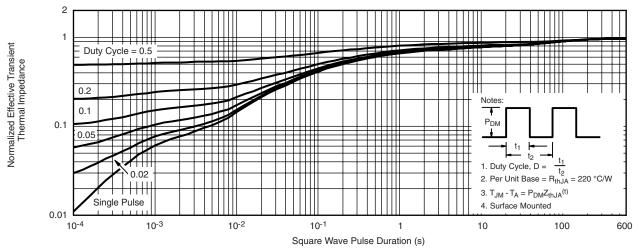




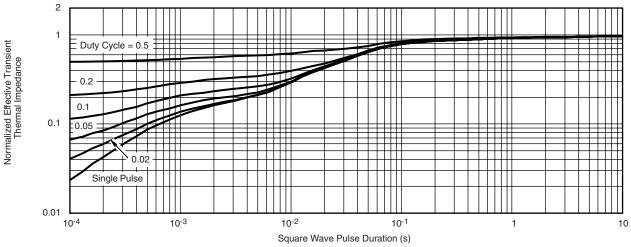
^{*} The power dissipation P_D is based on $T_{J(max)} = 150$ °C, using junction-to-case thermal resistance, and is more useful in settling the upper dissipation limit for cases where additional heatsinking is used. It is used to determine the current rating, when this rating falls below the package limit.

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TYPICAL CHARACTERISTICS 25 °C, unless otherwise noted



Normalized Thermal Transient Impedance, Junction-to-Ambient



Normalized Thermal Transient Impedance, Junction-to-Foot

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SC-70: 6-LEADS





	MILLIMETERS INCHES			S		
Dim	Min	Nom	Max	Min	Nom	Max
Α	0.90	-	1.10	0.035	_	0.043
A ₁	-	-	0.10	-	-	0.004
A_2	0.80	-	1.00	0.031	-	0.039
b	0.15	-	0.30	0.006	_	0.012
С	0.10	-	0.25	0.004	_	0.010
D	1.80	2.00	2.20	0.071	0.079	0.087
Ε	1.80	2.10	2.40	0.071	0.083	0.094
E ₁	1.15	1.25	1.35	0.045	0.049	0.053
е		0.65BSC			0.026BSC	;
e ₁	1.20	1.30	1.40	0.047	0.051	0.055
L	0.10	0.20	0.30	0.004	0.008	0.012
9	7°Nom 7°Nom					



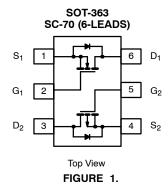
Dual-Channel LITTLE FOOT® 6-Pin SC-70 MOSFET Copper Leadframe Version Recommended Pad Pattern and Thermal Performance

INTRODUCTION

The new dual 6-pin SC-70 package with a copper leadframe enables improved on-resistance values and enhanced thermal performance as compared to the existing 3-pin and 6-pin packages with Alloy 42 leadframes. These devices are intended for small to medium load applications where a miniaturized package is required. Devices in this package come in a range of on-resistance values, in n-channel and p-channel versions. This technical note discusses pin-outs, package outlines, pad patterns, evaluation board layout, and thermal performance for the dual-channel version.

PIN-OUT

Figure 1 shows the pin-out description and Pin 1 identification for the dual-channel SC-70 device in the 6-pin configuration. Both n-and p-channel devices are available in this package — the drawing example below illustrates the p-channel device.



For package dimensions see outline drawing SC-70 (6-Leads) (http://www.vishay.com/doc?71154)

BASIC PAD PATTERNS

See Application Note 826, Recommended Minimum Pad Patterns With Outline Drawing Access for Vishay Siliconix MOSFETs, (http://www.vishay.com/doc?72286) for the SC-70 6-pin basic pad layout and dimensions. This pad pattern is sufficient for the low-power applications for which this package is intended. Increasing the drain pad pattern (Figure 2) yields a reduction in thermal resistance and is a preferred footprint.

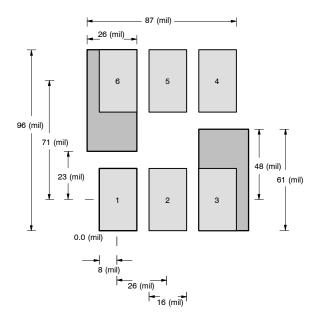


FIGURE 2. SC-70 (6 leads) Dual

EVALUATION BOARD FOR THE DUAL-CHANNEL SC70-6

The 6-pin SC-70 evaluation board (EVB) shown in Figure 3 measures 0.6 in. by 0.5 in. The copper pad traces are the same as described in the previous section, *Basic Pad Patterns*. The board allows for examination from the outer pins to the 6-pin DIP connections, permitting test sockets to be used in evaluation testing.

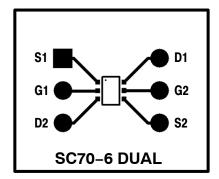
The thermal performance of the dual 6-pin SC-70 has been measured on the EVB, comparing both the copper and Alloy 42 leadframes. This test was then repeated using the 1-inch² PCB with dual-side copper coating.

A helpful way of displaying the thermal performance of the 6-pin SC-70 dual copper leadframe is to compare it to the traditional Alloy 42 version.

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Front of Board SC70-6



Back of Board SC70-6

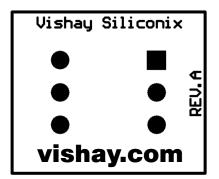


FIGURE 3.

THERMAL PERFORMANCE

Junction-to-Foot Thermal Resistance (the Package Performance)

Thermal performance for the dual SC-70 6-pin package is measured as junction-to-foot thermal resistance, in which the "foot" is the drain lead of the device as it connects with the body. The junction-to-foot thermal resistance for this device is typically 80°C/W, with a maximum thermal resistance of approximately 100°C/W. This data compares favorably with another compact, dual-channel package - the dual TSOP-6 which features a typical thermal resistance of 75°C/W and a maximum of 90°C/W.

Power Dissipation

The typical $R\theta_{JA}$ for the dual-channel 6-pin SC-70 with a copper leadframe is 224°C/W steady-state, compared to 413°C/W for the Alloy 42 version. All figures are based on the 1-inch² FR4 test board. The following example shows how the thermal resistance impacts power dissipation for the dual 6-pin SC-70 package at varying ambient temperatures.

Alloy 42 Leadframe

ALLOY 42 LEADFRAME				
Room Ambient 25 °C Elevated Ambient 60 °C				
$P_D = \frac{T_{J(max)} - T_A}{R\theta_{JA}}$	$P_D = \frac{T_{J(max)} - T_A}{R\theta_{JA}}$			
$P_{D} = \frac{150^{\circ}C - 25^{\circ}C}{413^{\circ}C/W}$	$P_{D} = \frac{150^{\circ}C - 60^{\circ}C}{413^{\circ}C/W}$			
$P_D = 303 \text{ mW}$	$P_D = 218 \text{ mW}$			

COOPER LEADFRAME				
Room Ambient 25 °C	Elevated Ambient 60 °C			
$P_D = \frac{T_{J(max)} - T_A}{R\theta_{JA}}$	$P_D = \frac{T_{J(max)} - T_A}{R\theta_{JA}}$			
$P_{D} = \frac{150^{\circ}C - 25^{\circ}C}{224^{\circ}C/W}$	$P_{D} = \frac{150^{\circ}C - 60^{\circ}C}{224^{\circ}C/W}$			
$P_D = 558 \text{ mW}$	$P_D = 402 \text{ mW}$			

Although they are intended for low-power applications, devices in the 6-pin SC-70 dual-channel configuration will handle power dissipation in excess of 0.5 W.

TESTING

To further aid the comparison of copper and Alloy 42 leadframes, Figures 4 and 5 illustrate the dual-channel 6-pin SC-70 thermal performance on two different board sizes and pad patterns. The measured steady-state values of $R\theta_{JA}$ for the dual 6-pin SC-70 with varying leadframes are as follows:

LITTLE FOOT 6-PIN SC-70					
Alloy 42 Copper					
Minimum recommended pad pattern on the EVB board (see Figure 3).	518°C/W	344°C/W			
Industry standard 1-inch ² PCB with maximum copper both sides.	413°C/W	224°C/W			

The results indicate that designers can reduce thermal resistance (θJA) by 34% simply by using the copper leadframe device as opposed to the Alloy 42 version. In this example, a 174°C/W reduction was achieved without an increase in board area. If an increase in board size is feasible, a further 120°C/W reduction can be obtained by utilizing a 1-inch². PCB area.

The Dual copper leadframe versions have the following suffix:

Dual: Si19xxEDH Si15xxEDH Compl.:

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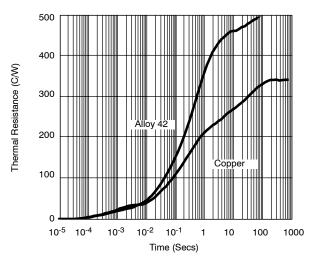


FIGURE 4. Dual SC70-6 Thermal Performance on EVB

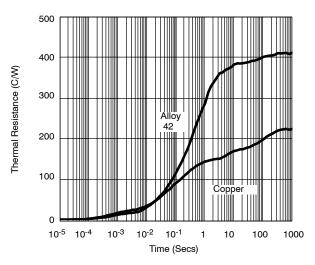


FIGURE 5. Dual SC70-6 Comparison on 1-inch² PCB



RECOMMENDED MINIMUM PADS FOR SC-70: 6-Lead



Recommended Minimum Pads Dimensions in Inches/(mm)

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